

Average Behavioral Modeling Technique for Switched- Capacitor Voltage Converters

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Mentor
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 - ▲ Average model analysis
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Introduction

- ▶ **Complex systems today are becoming more and more mixed-signal, with the analog part being the design and verification bottleneck**
- ▶ **Traditionally, the virtual verification for the analog part of the design was only available through transistor-level SPICE simulations.**
- ▶ **Provide accurate results BUT require extensive computations**
- ▶ **Very often, speed is more critical than accuracy**

Introduction Cont.

- ▶ **Recently, behavioral modeling has been proved to be the right methodology to cope with today's design and verification demands**
- ▶ **However, some behavioral models of complex systems still consume a lot of simulation time**
- ▶ **May be caused by high switching rates**
 - ▲ **Simulation time-step is bound by the switching period**
- ▶ **Actual information of interest is usually of a much lower frequency**

Average Modeling Approach

- ▶ **Average models concentrate on the information bearing signal of lower frequency**
 - ▲ Simulation time-step is greatly relaxed
- ▶ **Detailed high frequency transients and ripples produced by the actual circuit are discarded**
- ▶ **Useful in:**
 - ▲ Efficient system level simulations,
 - ▲ Design parameter exploration,
 - ▲ Obtaining engineering intuition into the operation of these switched circuits

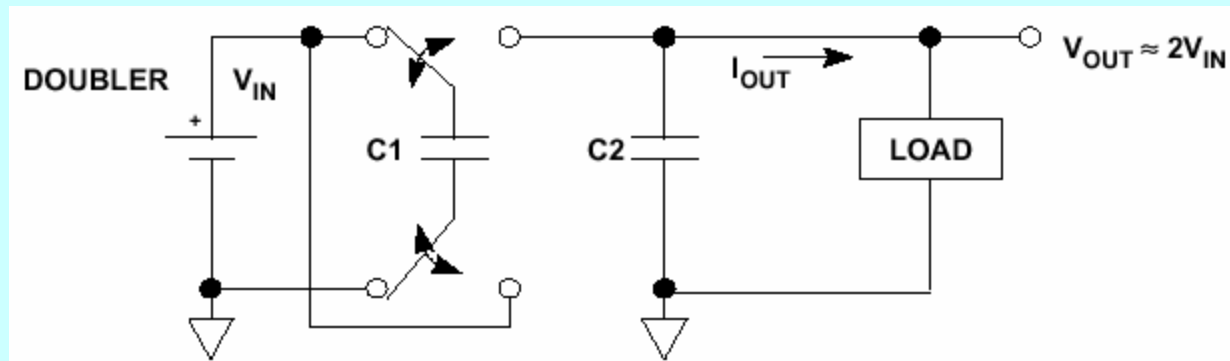
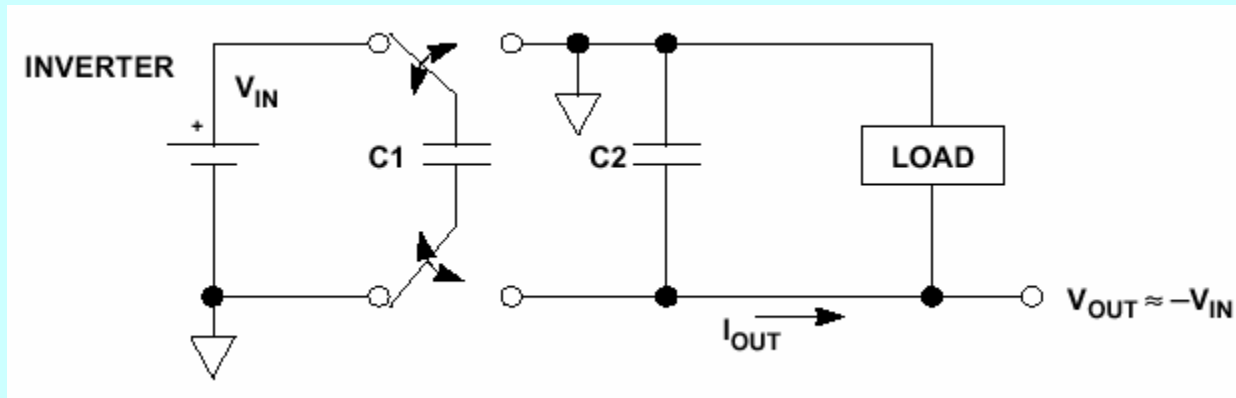
Switched Capacitor DC-DC Converters

- ▶ Also known as charge pump DC-DC converters
- ▶ Used for:
 - ▶ Multiplying the voltage level available from a low-voltage battery
 - ▶ Generate invert voltages
- ▶ Accomplish energy transfer and voltage conversion using capacitors and semiconductor switches – *Inductorless*

Switched Capacitor DC-DC Converters

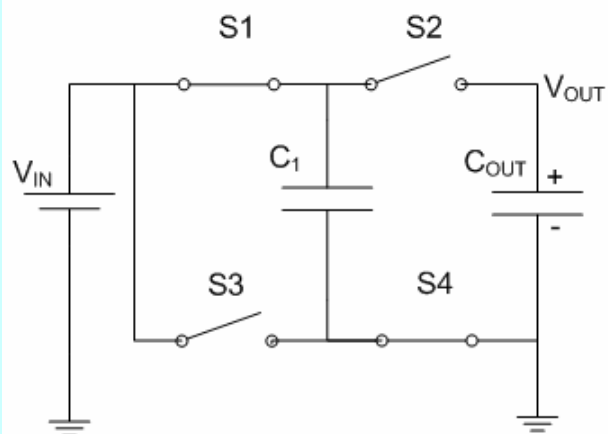
- ▶ **Advantages:**
 - ▶ Do not require magnetic components
 - ▶ Simple, small, low cost
- ▶ **Recent trend on low-power-low-voltage circuit design and applications of portable equipments leads to renewed interest on charge pump circuits**
- ▶ **Mandatory in power management ICs for battery powered portable applications**

Types of Switched-Capacitor DC-DC Converter Circuits



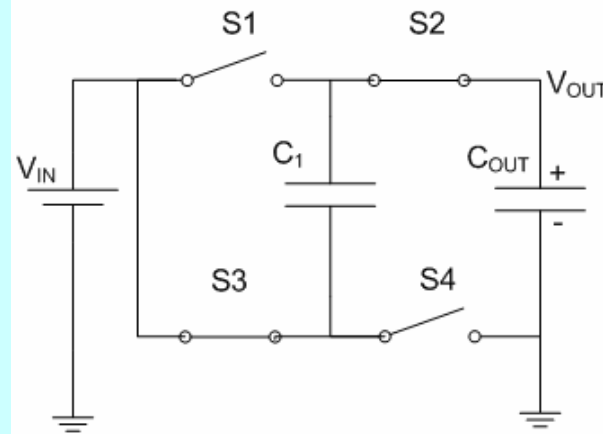
Voltage Doubler Basic Principle of Operation

ON State:



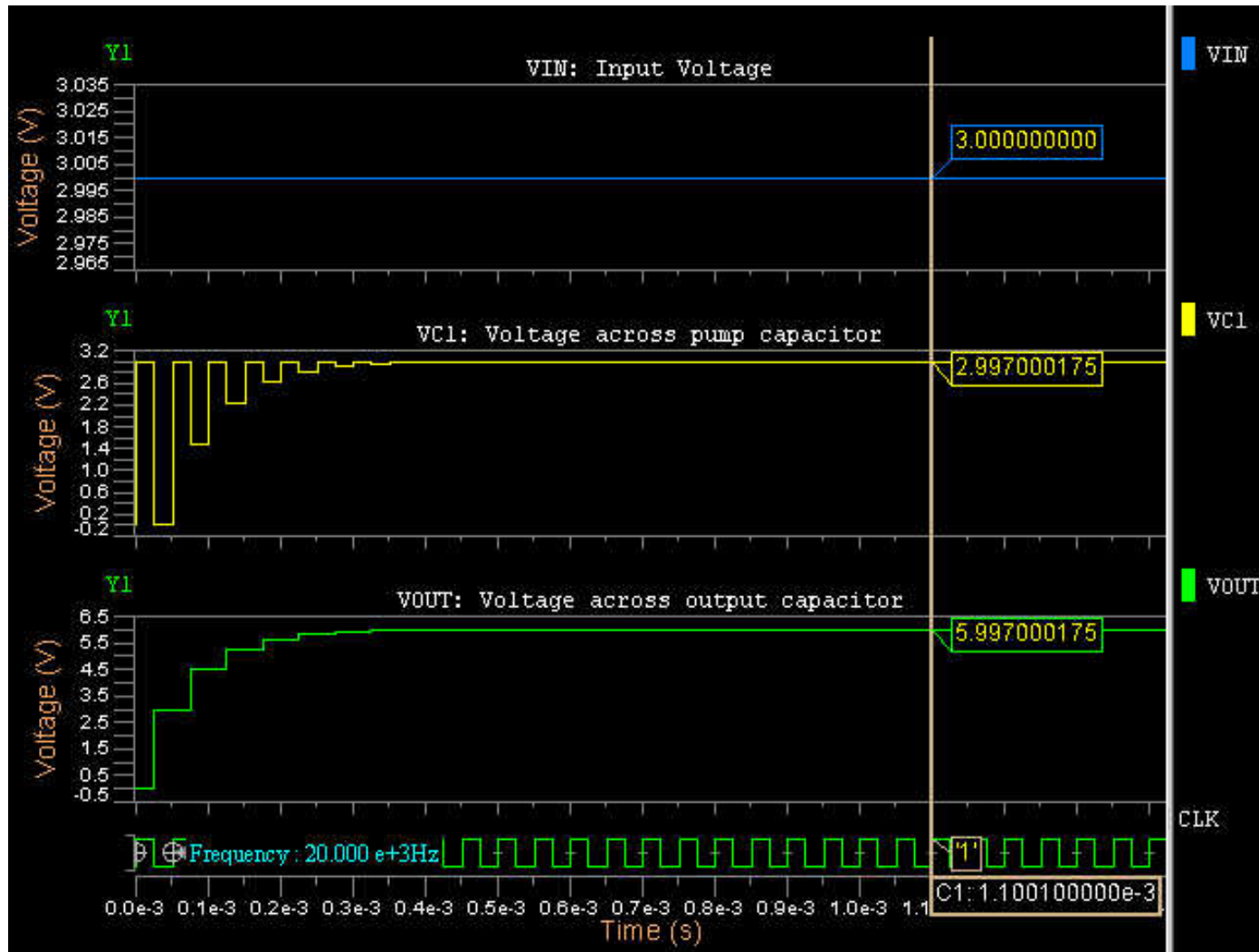
- Connect C_1 (pump capacitor) in parallel with V_{IN}
- Charge C_1 by V_{IN}
- C_{OUT} discharges through load

OFF State:



- Connect C_1 in series with V_{IN} and C_{OUT}
- Compensate voltage loss in C_{OUT} (charge sharing)
- Exponential growth targeting $(2V_{IN})$

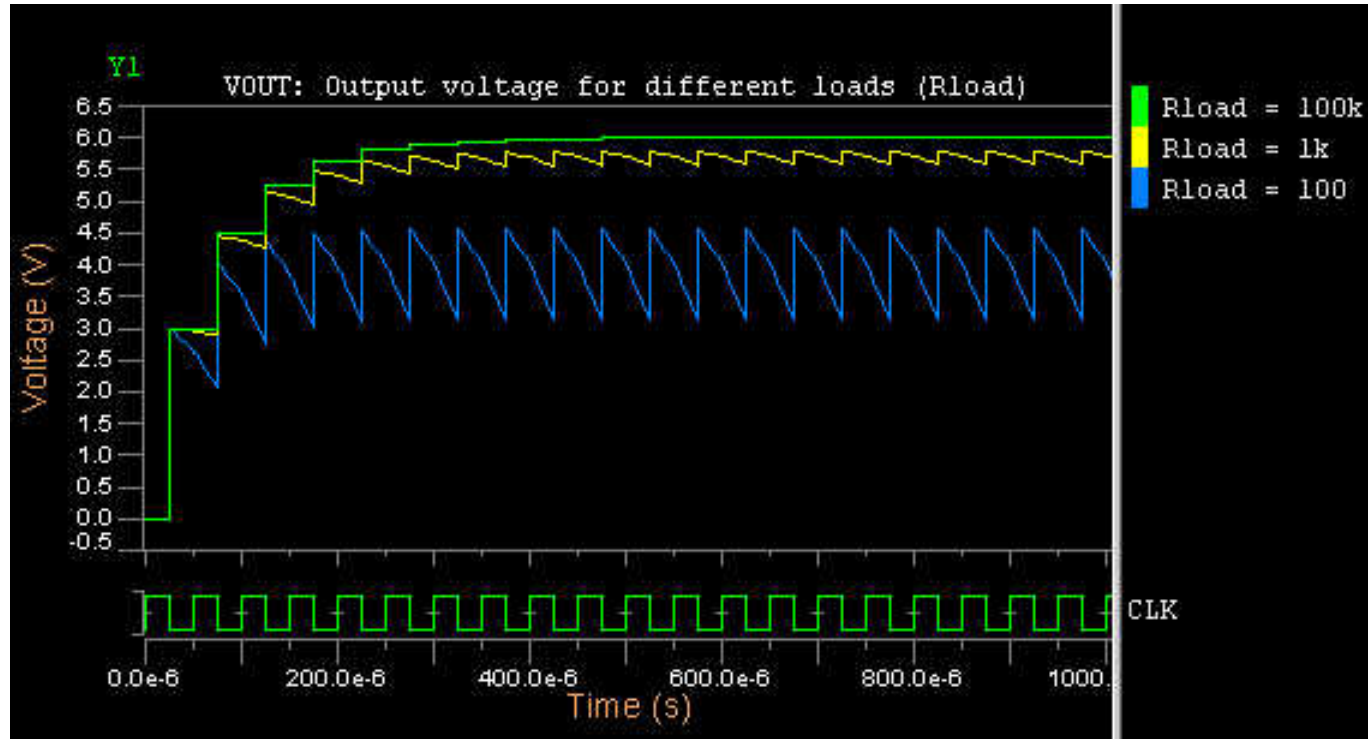
Ideal Voltage Doubler Spice Simulation



$f = 20\text{kHz}$,
 $C_1 = 1\mu\text{F}$,
 $C_{\text{OUT}} = 1\mu\text{F}$,
 $R_{\text{LOAD}} = 100\text{k}$

Assume ideal switches and capacitors

Ideal Voltage Doubler Spice Simulation *with varying output load*

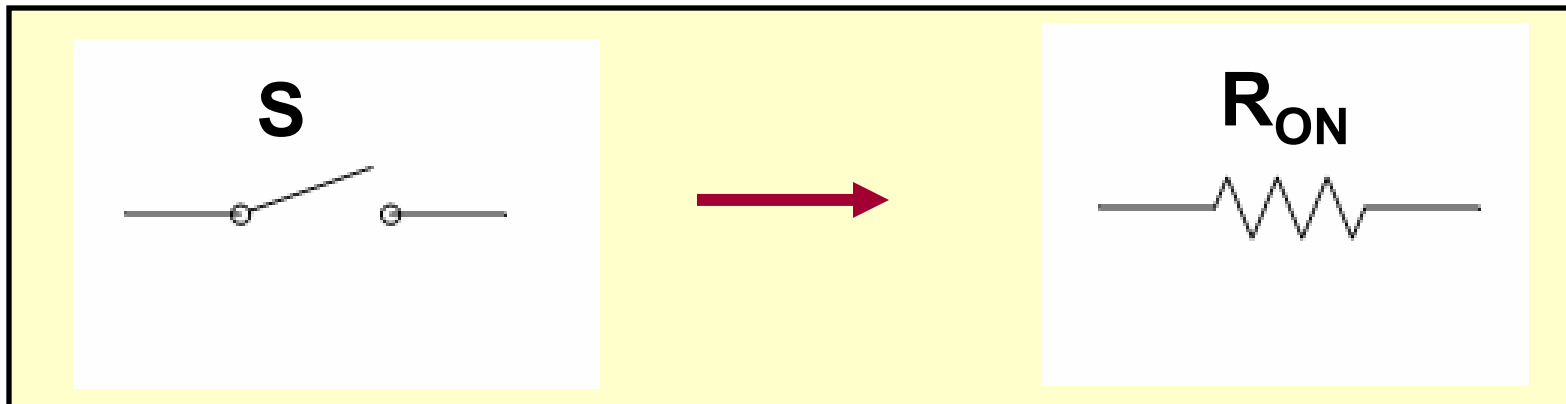


$f = 20\text{kHz}$,
 $C_1 = 1\mu\text{F}$,
 $C_{\text{OUT}} = 1\mu\text{F}$,
 $R_{\text{LOAD}} = 100, 1\text{k}, 100\text{k}$

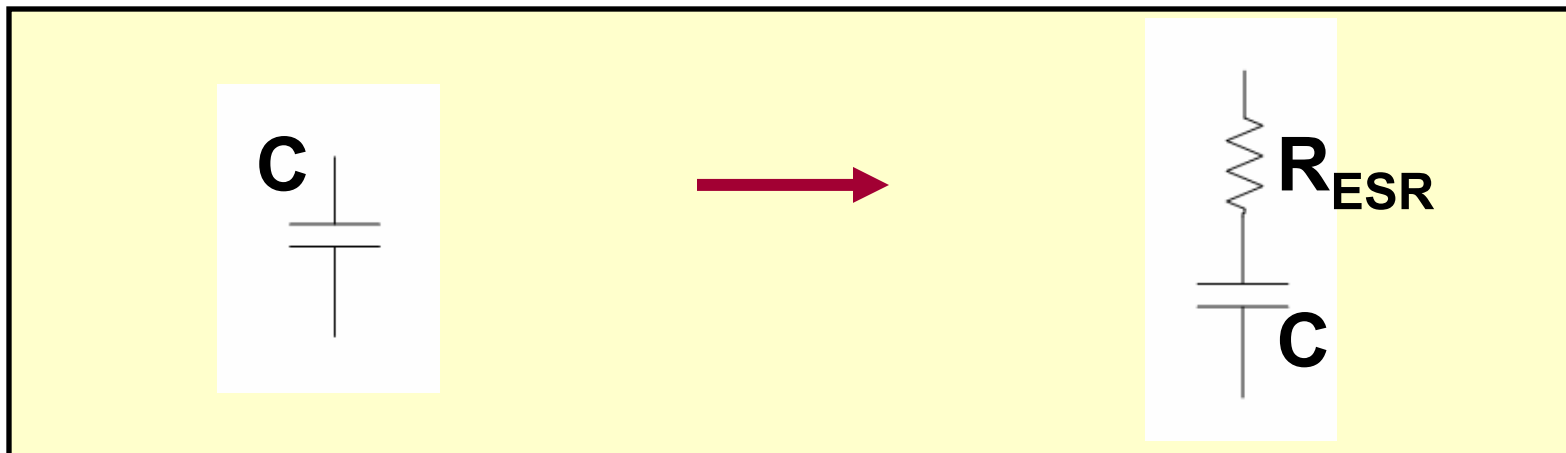
Assume ideal switches and capacitors

Non-Ideal Voltage Doubler

► Non-ideal switch



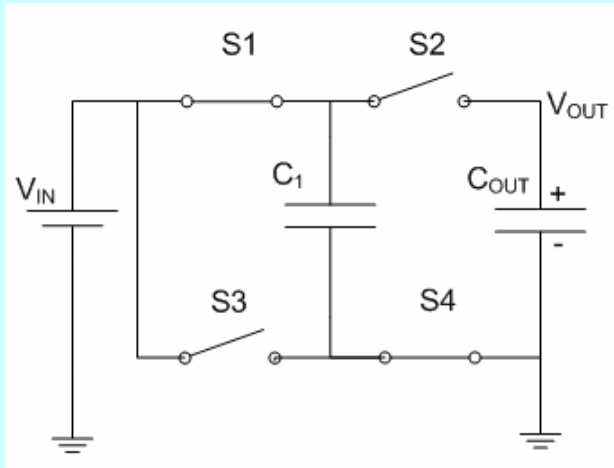
► Non-ideal capacitor



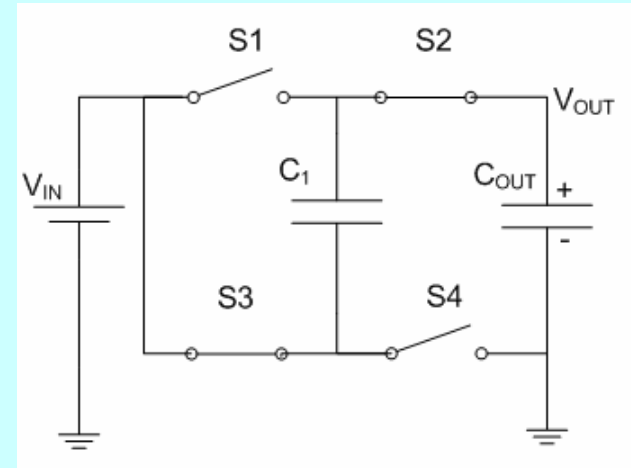
Non-Ideal Voltage Doubler

Ideal

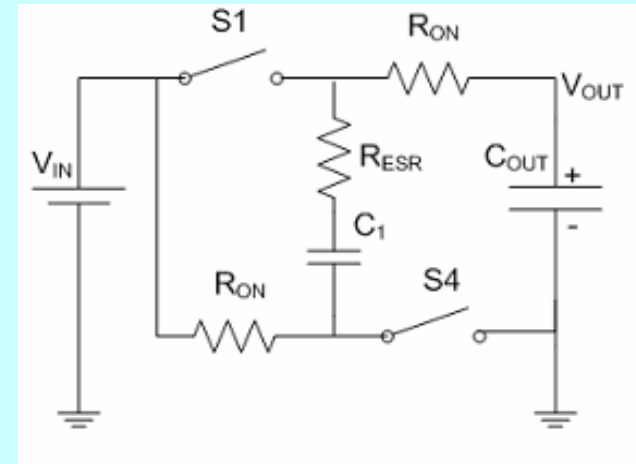
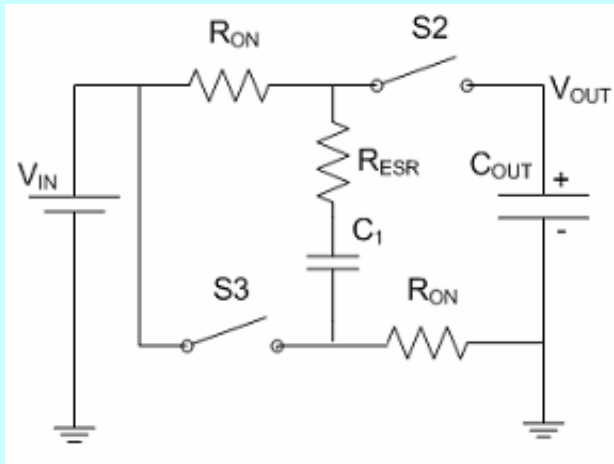
ON State:



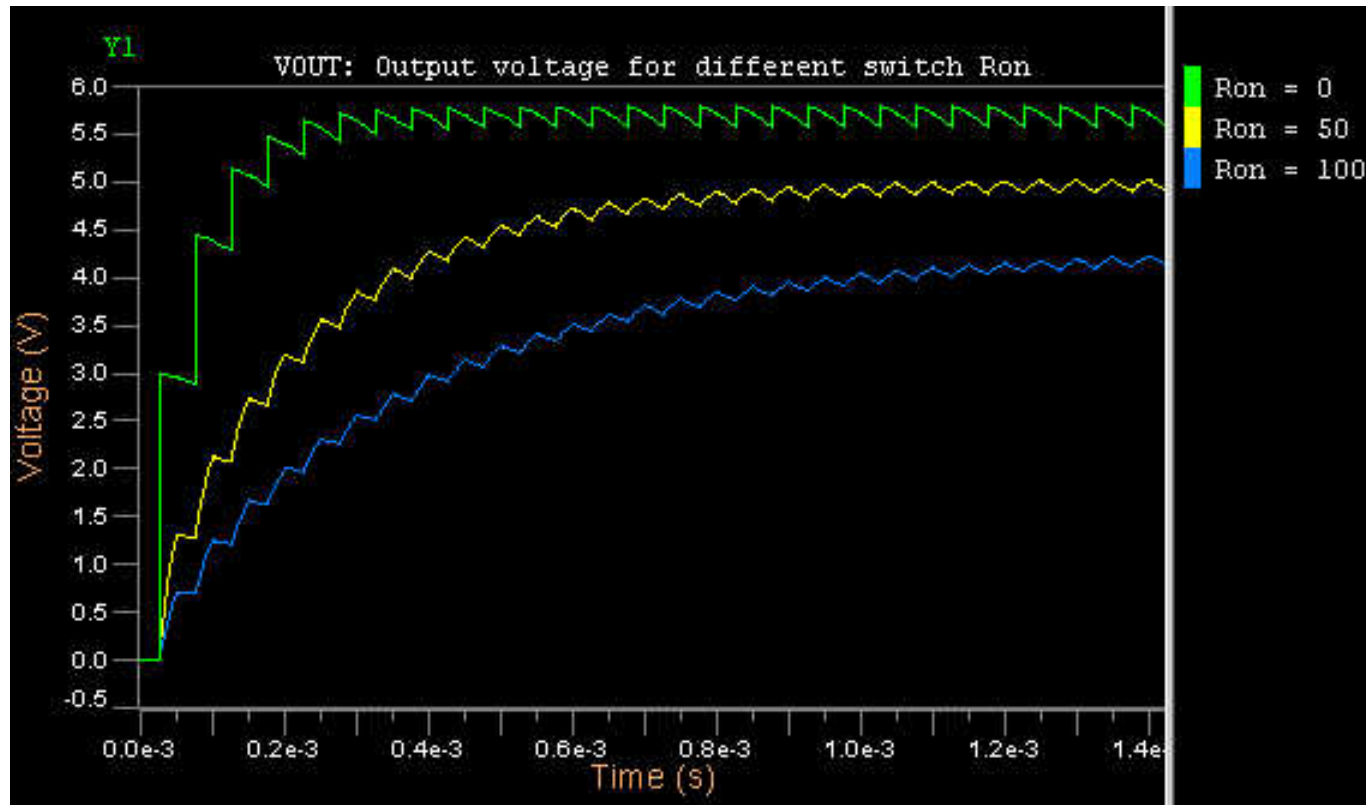
OFF State:



Non-Ideal
(R_{ON} , R_{ESR})



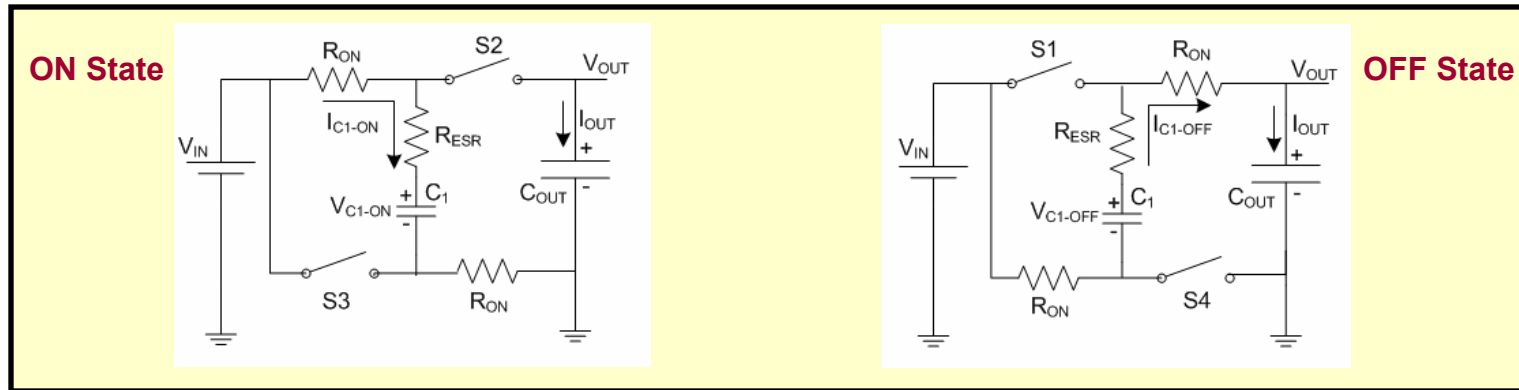
Non-Ideal Voltage Doubler Spice Simulation



$f = 20\text{kHz}$,
 $C_1 = 1\mu\text{F}$,
 $C_{\text{OUT}} = 1\mu\text{F}$,
 $R_{\text{LOAD}} = 100\text{k}$,
 $R_{\text{ESR}} = 0$,
 $R_{\text{ON}} = 0, 50, 100$

Assume non-ideal switches (R_{ON})

Average Model: Analysis



ON STATE:

$$V_{C1-ON}(t) = V_{IN} - 2R_{ON} I_{C1-ON}(t) - R_{ESR} I_{C1-ON}(t)$$

Average voltage during ON state:

$$V_{C1-ON-AVE} = V_{IN} - 2R_{ON} \cdot I_{C1-ON-AVE} - R_{ESR} I_{C1-ON-AVE} \quad \dots (1)$$

OFF STATE:

$$V_{C1-OFF}(t) = V_{OUT}(t) - V_{IN} + 2R_{ON} I_{C1-OFF}(t) - R_{ESR} I_{C1-OFF}(t)$$

Average voltage during ON state:

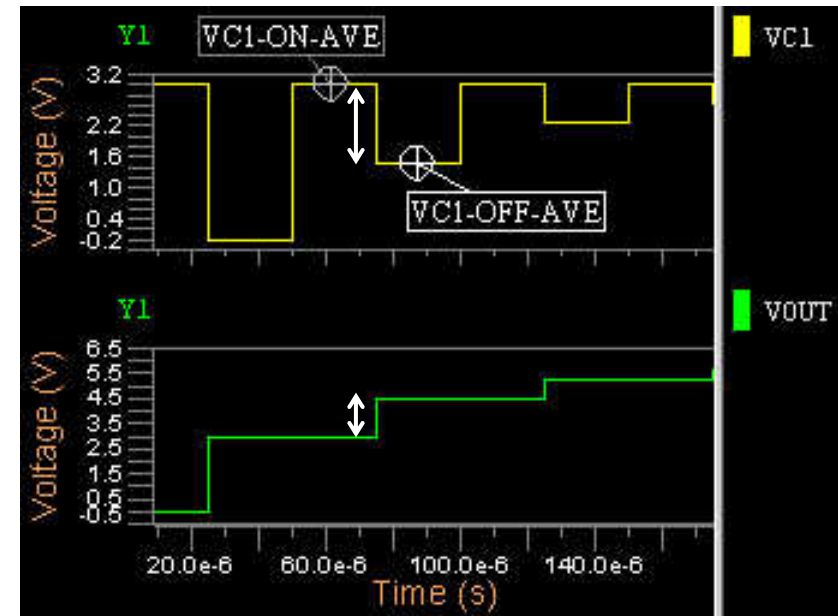
$$V_{C1-OFF-AVE} = V_{OUT-AVE} - V_{IN} + 2R_{ON} I_{C1-OFF-AVE} - R_{ESR} I_{C1-OFF-AVE} \quad \dots (2)$$

Difference charge stored in C_1 between ON and OFF states

=

Net charge/current transferred to the output in one cycle:

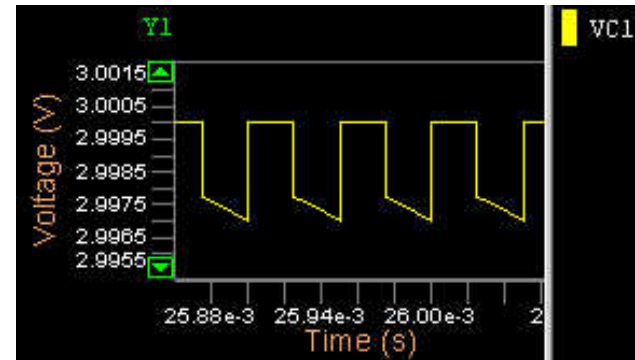
$$\begin{aligned}
 I_{\text{OUT-AVE}} &= f \cdot \Delta Q_{C1\text{-AVE}} \\
 &= f \cdot (Q_{C1\text{-ON-AVE}} - Q_{C1\text{-OFF-AVE}}) \\
 &= f C_1 (V_{C1\text{-ON-AVE}} - V_{C1\text{-OFF-AVE}})
 \end{aligned}$$



Using (1) and (2):

$$\begin{aligned}
 I_{\text{OUT-AVE}} &= f C_1 \{ 2V_{\text{IN}} - V_{\text{OUT-AVE}} \\
 &\quad - (2R_{\text{ON}} + R_{\text{ESR}}) I_{C1\text{-ON-AVE}} - (2R_{\text{ON}} + R_{\text{ESR}}) I_{C1\text{-OFF-AVE}} \} \quad \dots (3)
 \end{aligned}$$

Due to law of conservation of charge in C_1 :



amount of charge flowing into C_1 during ON state at S.S. (ΔQ_{ON})

=

amount of charge flowing out of C_1 at OFF state at S.S. (ΔQ_{OFF})

$$\begin{aligned} \Delta Q_{C1-ON-AVE} &= \Delta Q_{C1-OFF-AVE} \\ I_{C1-ON-AVE} \cdot DT &= I_{C1-OFF-AVE} \cdot (1-D)T \\ I_{C1-ON-AVE} \cdot D &= I_{C1-OFF-AVE} \cdot (1-D) \end{aligned} \quad \dots (4)$$

Also, the average output current:

$$I_{OUT-AVE} = I_{C1-OFF-AVE} \cdot (1-D)$$

Therefore,

$$I_{C1-OFF-AVE} = I_{OUT-AVE} / (1-D) \quad \dots (5)$$

$$I_{C1-ON-AVE} = I_{OUT-AVE} / D \quad \dots (6)$$

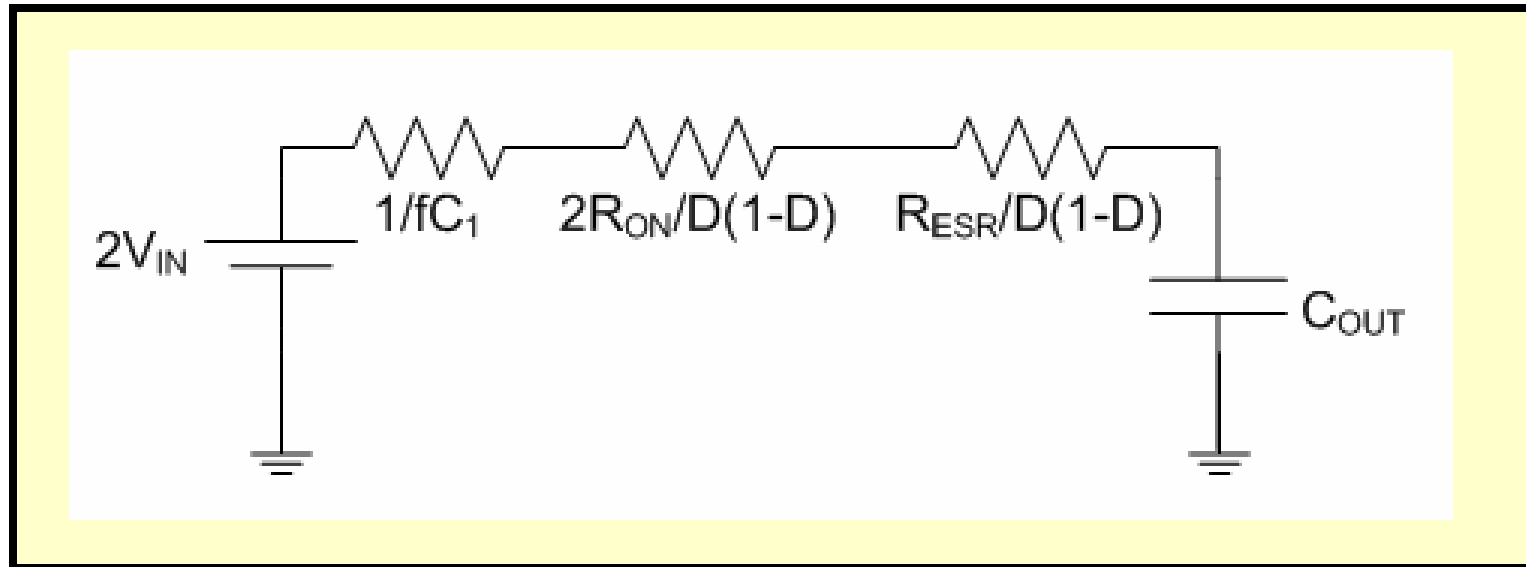
Substitute (5) & (6) in (3):

$$I_{\text{OUT-AVE}} = fC_1 \cdot \{ 2V_{\text{IN}} - V_{\text{OUT-AVE}} - (2R_{\text{ON}} + R_{\text{ESR}}) \cdot (I_{\text{OUT-AVE}} / (D(1-D))) \}$$

Which can be re-written as,

$$2V_{\text{IN}} - V_{\text{OUT-AVE}} = \left\{ \frac{1}{fC_1} + \frac{2R_{\text{ON}} + R_{\text{ESR}}}{D(1-D)} \right\} \cdot I_{\text{OUT-AVE}}$$

Average Equivalent Circuit



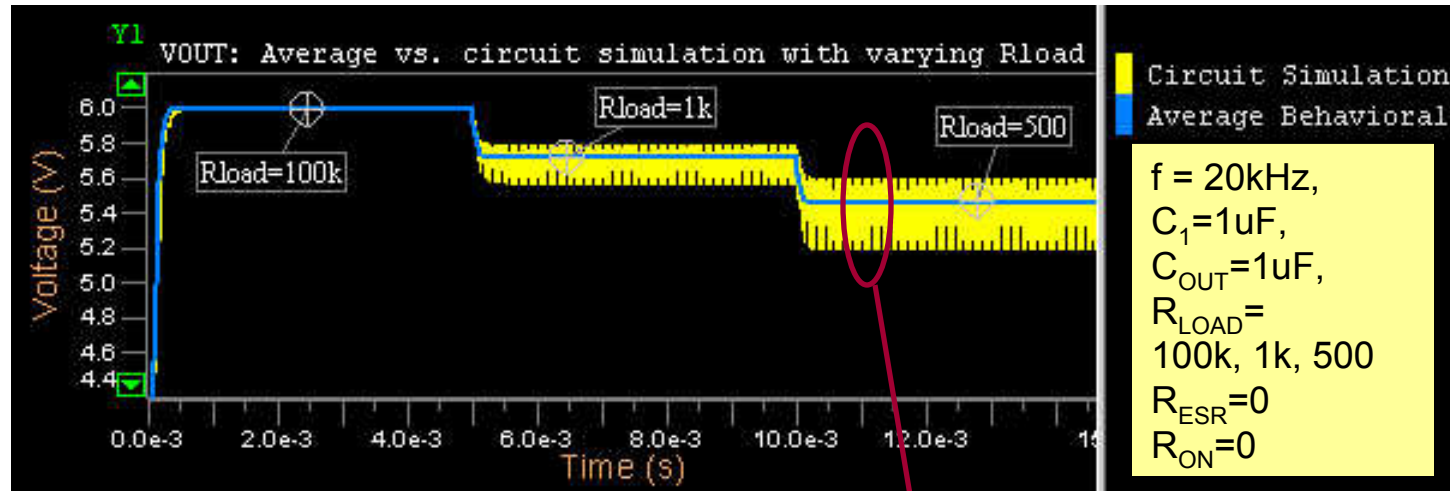
Required model parameters (given by user) :

C_1	:	<i>Pump Capacitance</i>
R_{ESR}	:	<i>Pump Capacitor Equivalent Series Resistance</i>
R_{ON}	:	<i>Switch ON Resistance</i>
F	:	<i>Switching Frequency</i>
D	:	<i>Switching Duty Cycle</i>

Experimental Setup

- ▶ **Circuit level simulation:**
 - ▲ Constructed using Eldo™ primitives (voltage sources, resistors, capacitors, switch macromodels)
 - ▲ Transient analysis
 - ▲ Varying output resistive load (R_{load}) parallel to C_{OUT}
 - ▲ Simulated using ADVance MS™
- ▶ **Behavioral Level Simulation (VHDL-AMS)**
 - ▲ Instanciated within Eldo™ netlist and connected to Eldo™ voltage source for input
 - ▲ Transient analysis
 - ▲ Varying output resistive load (R_{load}) parallel to C_{OUT}
 - ▲ Simulated using ADVance MS™

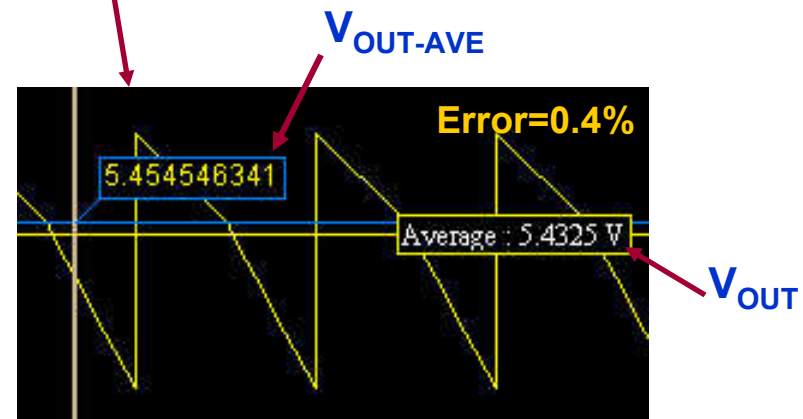
Experimental Results: Waveforms



$$\text{Error}(\%) = \frac{V_{\text{OUT}} - V_{\text{OUT-AVE}}}{V_{\text{OUT}}} \cdot 100\%$$

Where,

$V_{\text{OUT-AVE}}$: average behavioral model output
 V_{OUT} : calculated average of the circuit simulation output during one period



Experimental Results: Statistics

▶ Accuracy:

- ▲ $R_{LOAD} = 100k$: $(5.9968 - 5.9970) / 5.9968 \times 100 = 0\% \text{ Error}$
- ▲ $R_{LOAD} = 1k$: $(5.6976 - 5.7143) / 5.6976 \times 100 = 0.3\% \text{ Error}$
- ▲ $R_{LOAD} = 500$: $(5.4325 - 5.4545) / 5.4325 \times 100 = 0.4\% \text{ Error}$

▶ Simulation Timepoints:

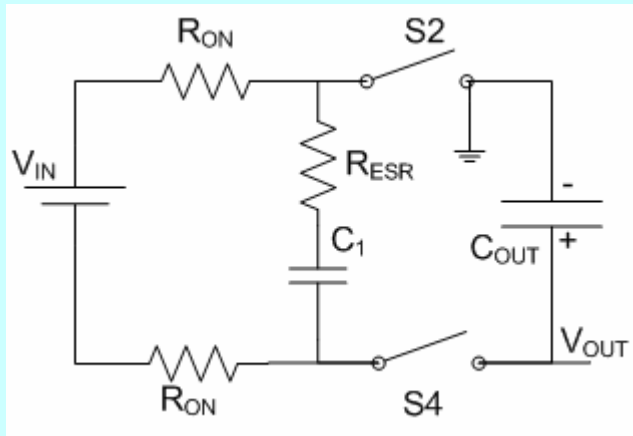
- ▲ Switching Circuit: 560066 points
- ▲ Average Model: 164 points
- ▲ Saving: > 99 %

▶ Speed:

- ▲ Switching Circuit: 1mn 34s 970ms
- ▲ Average Model: 0s 020ms
- ▲ Speed Gain: 4750 X

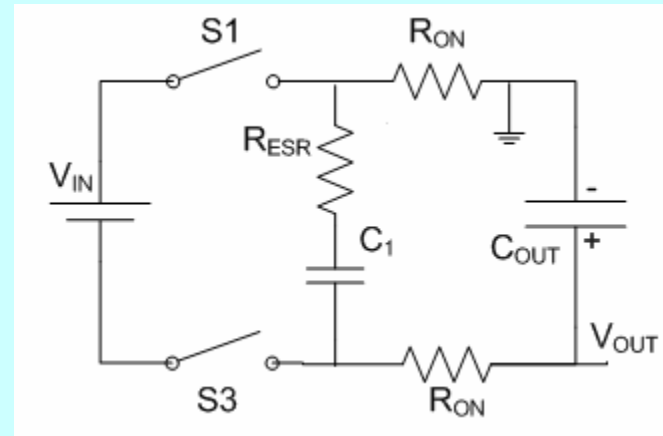
Voltage Inverter Basic Principle of Operation

ON State:



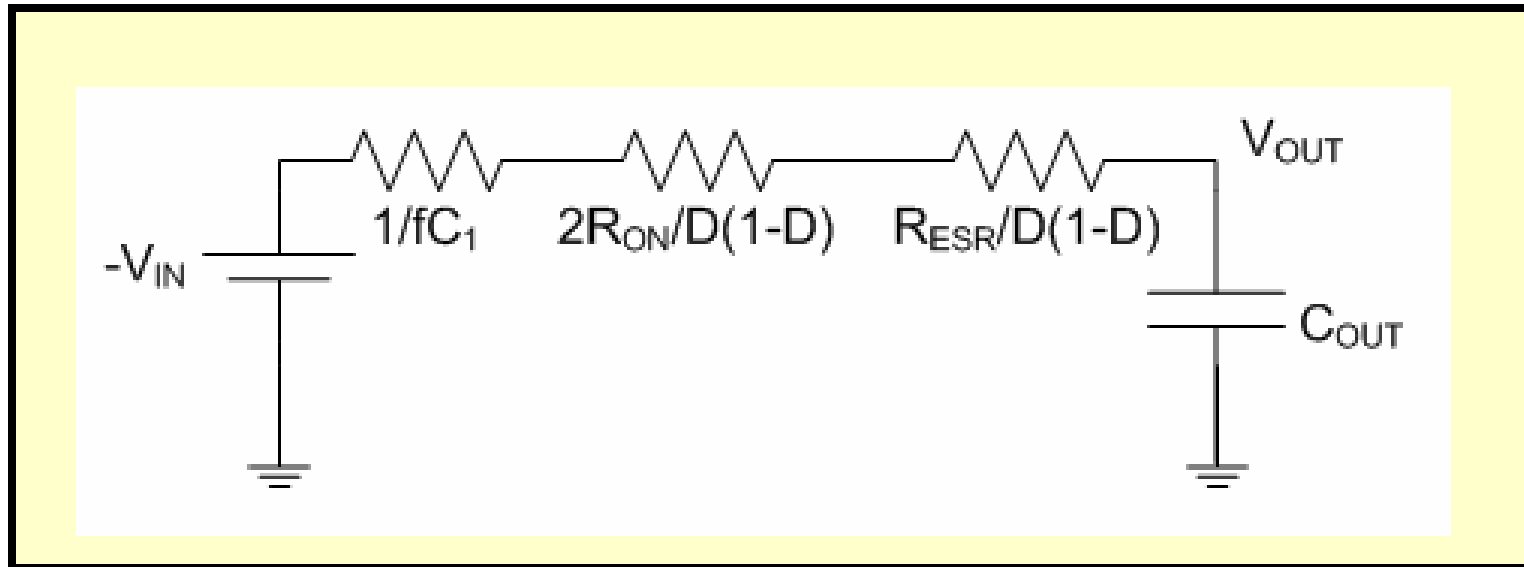
- Connect C_1 (pump capacitor) in parallel with V_{IN}
- Charge C_1 by V_{IN}
- C_{OUT} discharges through load.

OFF State:



- Connect C_1 in parallel with C_{OUT}
- Compensate voltage loss in C_{OUT} (charge sharing)
- Exponential growth targeting ($-V_{IN}$)

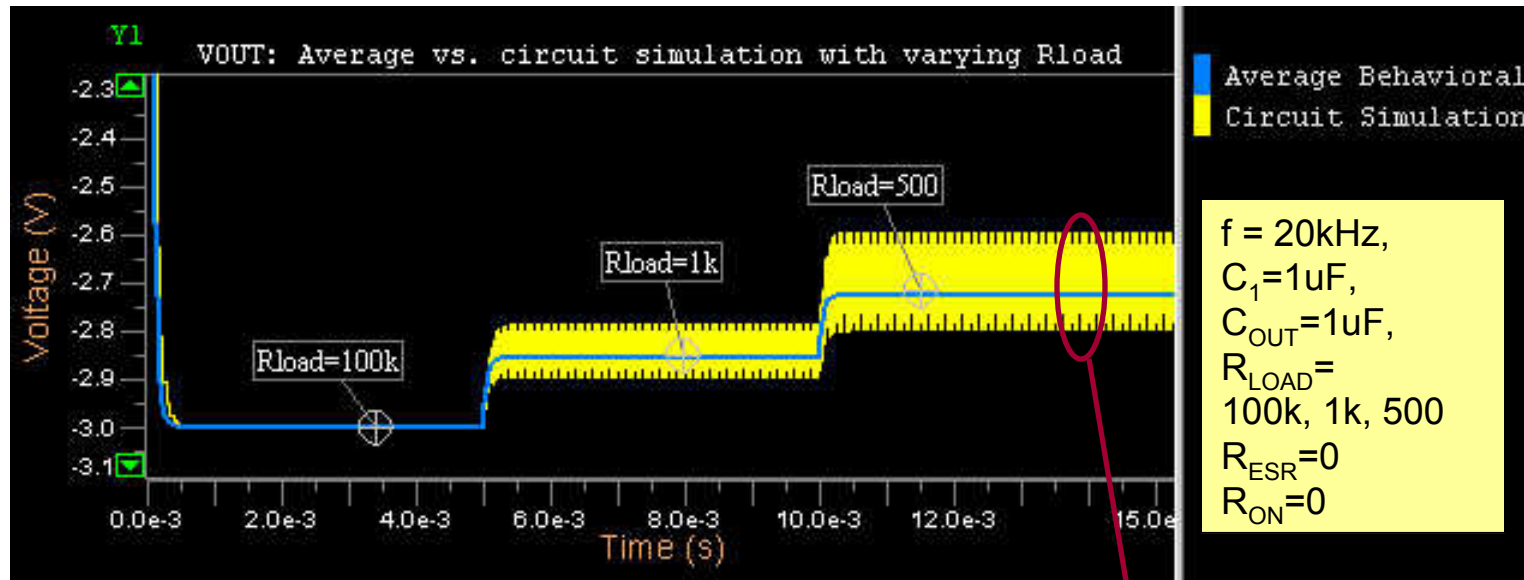
Average Equivalent Circuit



Required model parameters (given by user) :

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R_{ON}	:	<i>Switch ON Resistance</i>
F	:	<i>Switching Frequency</i>
D	:	<i>Switching Duty Cycle</i>

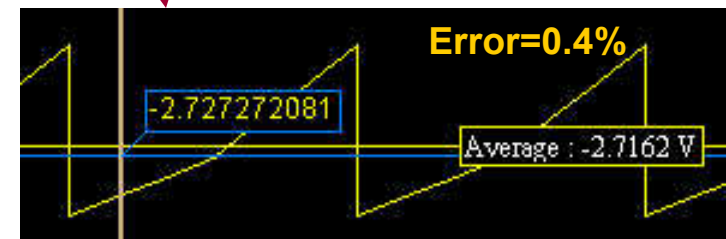
Experimental Results



Accuracy: 0 - 0.4% Error

Timepoints: 560066 vs. 163 point (>99% saving)

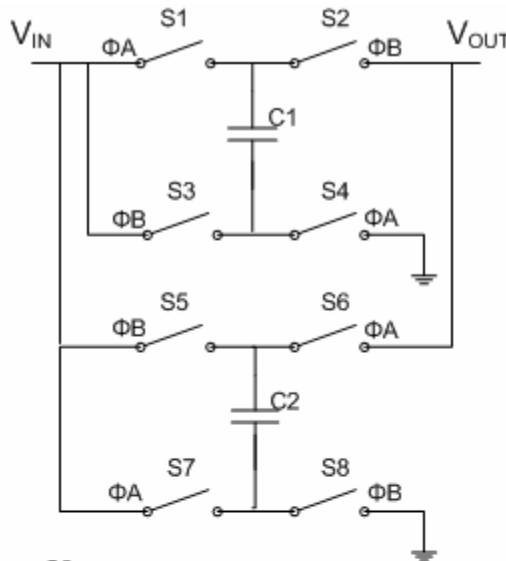
Speed: 1mn 37s 930ms vs. 20ms (4900 X faster)



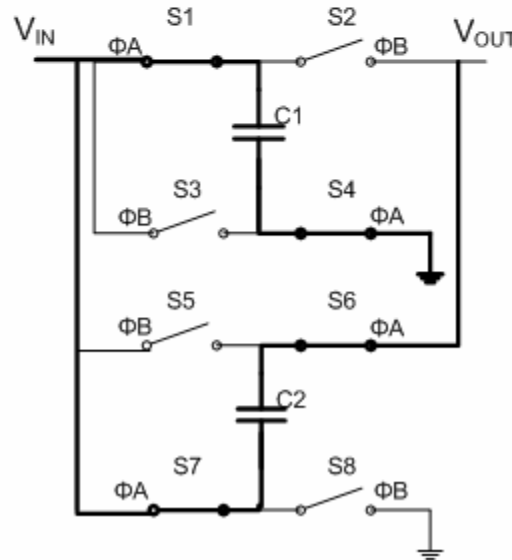
Voltage Doubler Push-Pull Configuration

- ▶ **Two voltage doublers run in parallel and in opposite phases**
 - ▲ **When one pump is being charged, the other is charging the output**
- ▶ **In this architecture, one of the pump capacitors is always delivering charge to the output**
- ▶ **Advantages:**
 - ▲ **Minimize voltage loss and output voltage ripple**
 - ▲ **Allows the use of smaller output capacitor compared to a conventional voltage doubler**

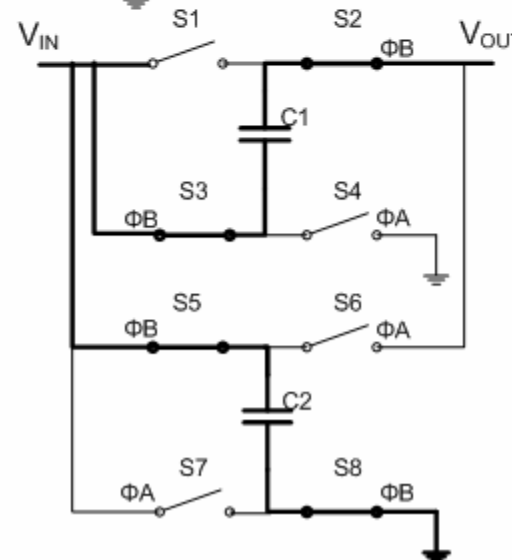
Push-Pull Configuration



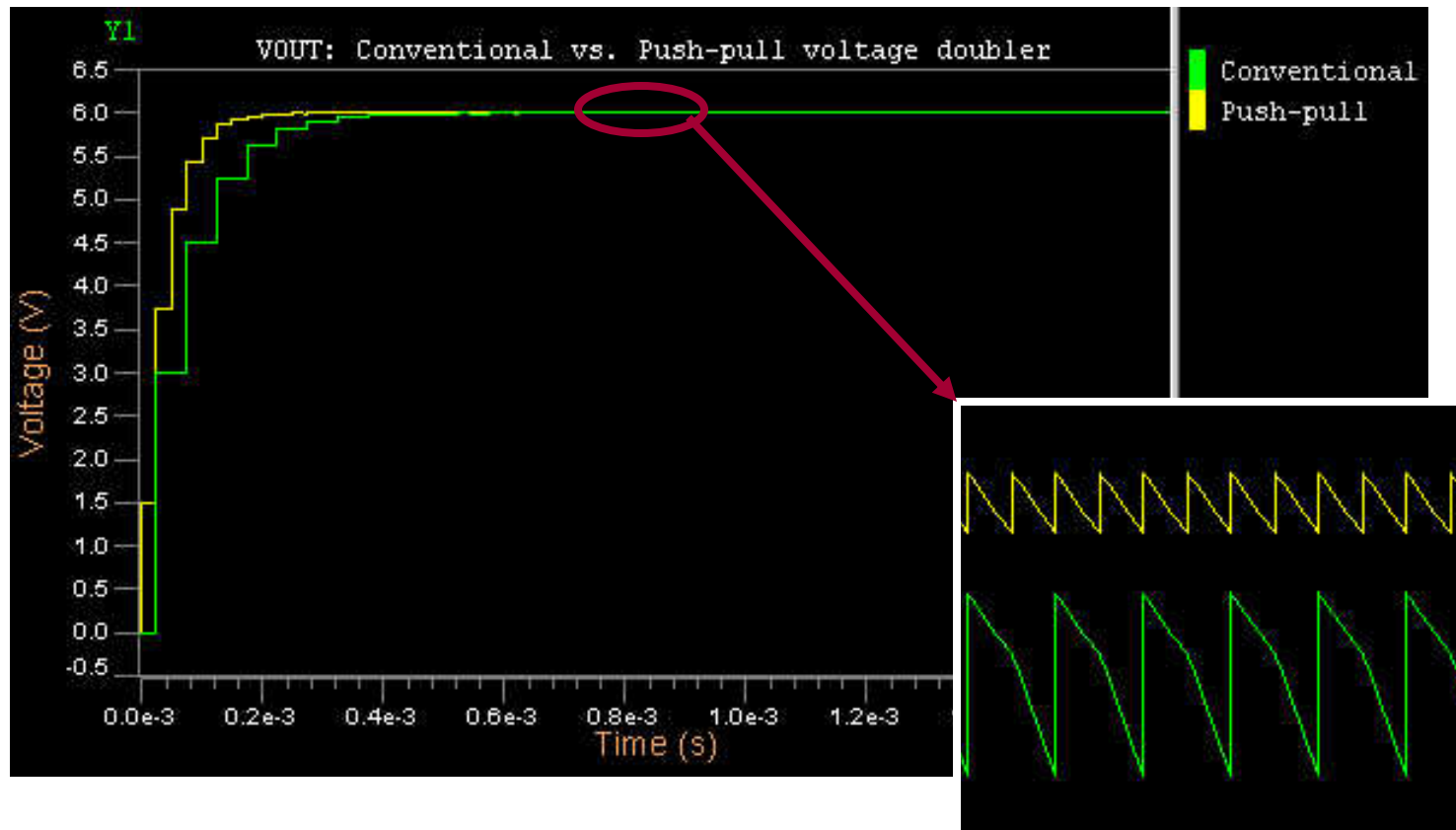
Phase I



Phase II

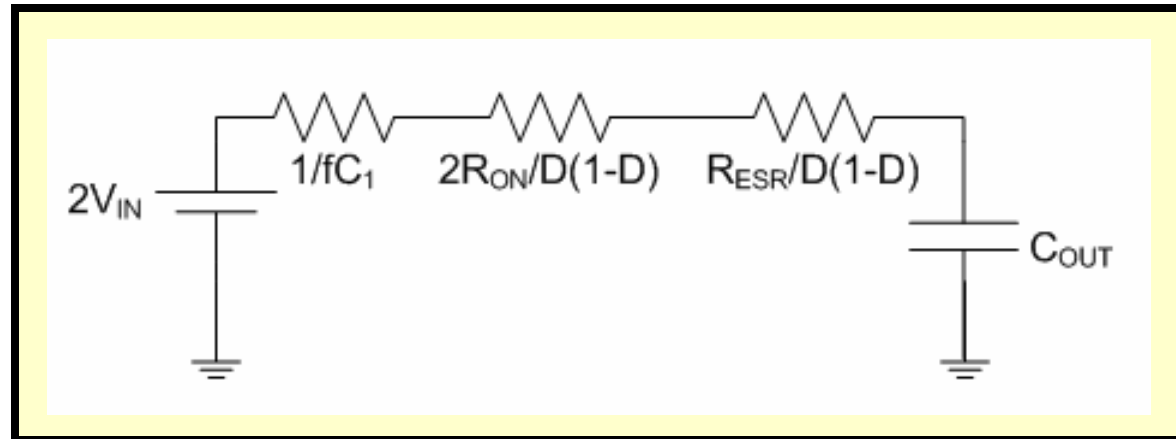


Conventional Voltage Doubler vs. Push-Pull Voltage Doubler

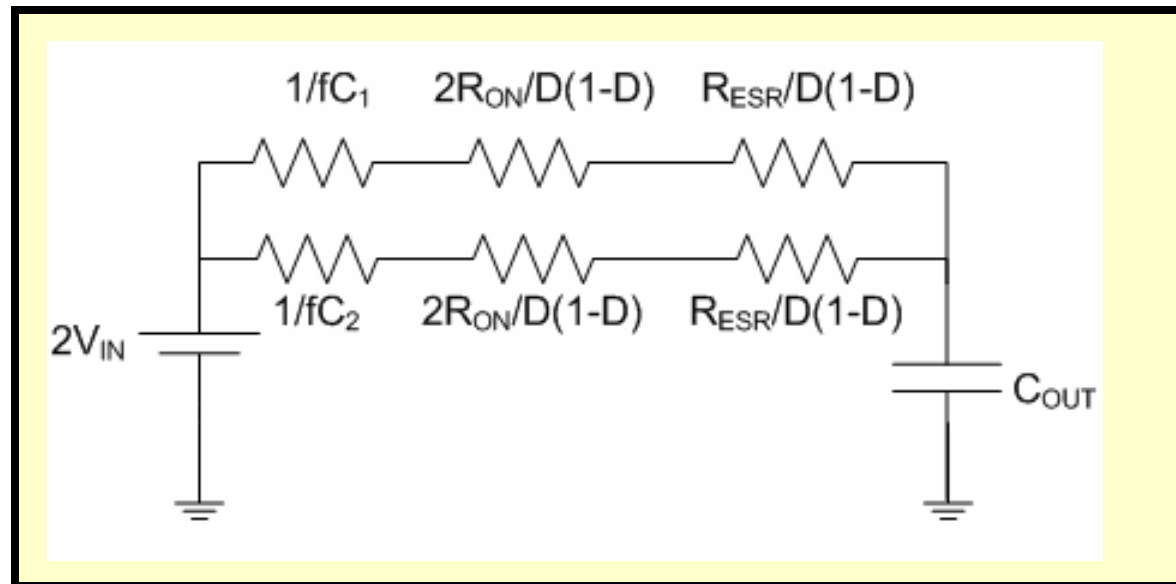


Modified Equivalent Average Circuit

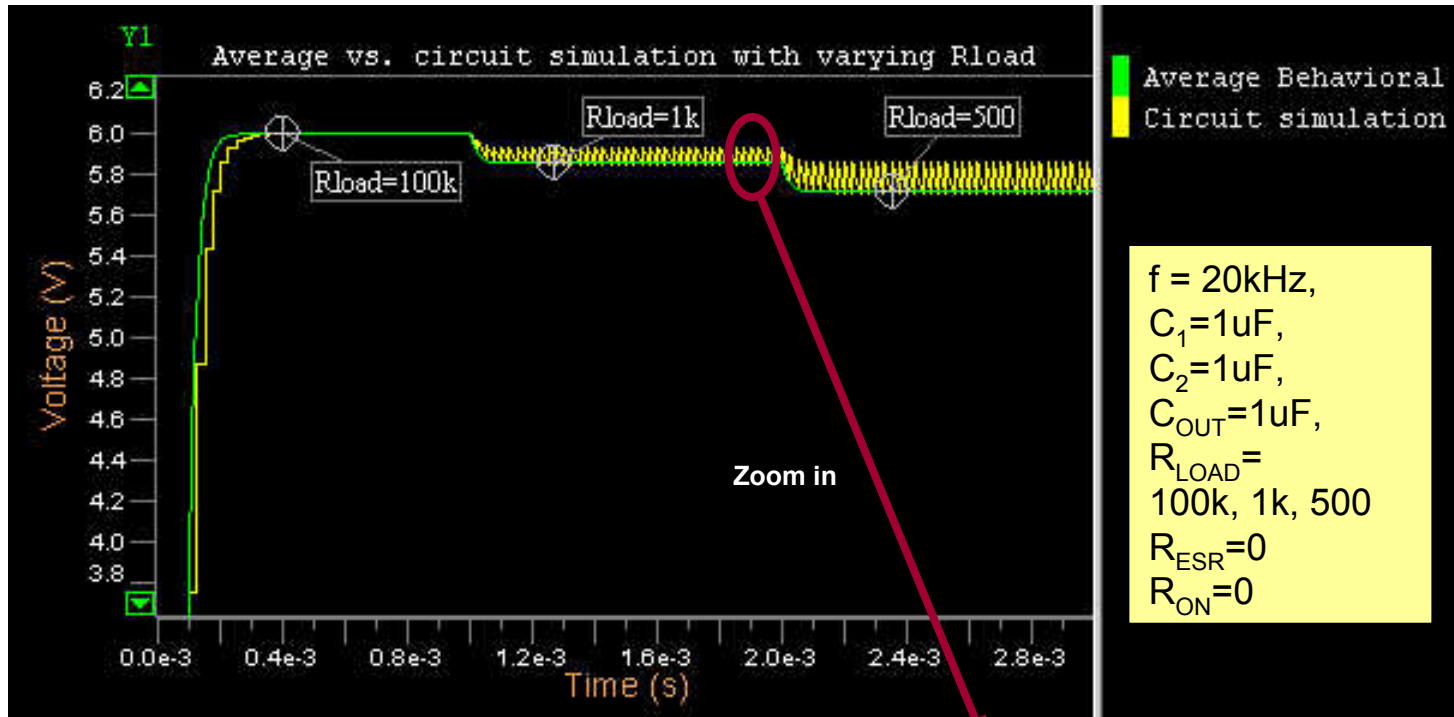
Conventional



Push Pull



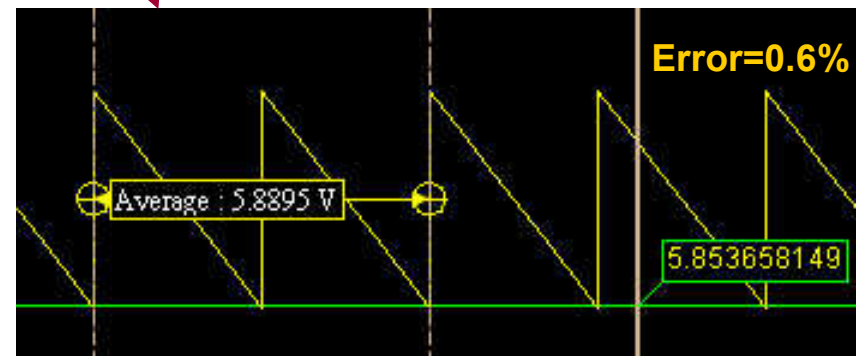
Experimental Results



Accuracy: 0.007 - 1.2% Error

Timepoints: 460066 vs. 156 point (>99% saving)

Speed: 1mn 43s 980ms vs. 20ms (5200 X faster)



Experimental Results (Error%)

f=20kHz

Rload	R_{ON}=0	R_{ON}=10	R_{ON}=50	R_{ON}=100
100k	0.007	0.0167	0.023	0.025
1k	0.6	1.6	1.9	1.9
500	1.2	2.3	3.7	3.3

f=250kHz

Rload	R_{ON}=0	R_{ON}=10	R_{ON}=50	R_{ON}=100
100k	0	0.0017	0.0017	0.0017
1k	0.05	0.19	0.18	0.17
500	0.1	0.37	0.31	0.28

Conclusion

- ▶ Average modeling is beneficial when the nature of the circuit includes 2 frequencies
 - ▲ High frequency → switching frequency
 - ▲ Lower frequency → information of interest
- ▶ Average modeling focuses on the lower frequency
- ▶ Discard the detailed analysis of high frequency component leading to large simulation speed gain
- ▶ Re-visited an average modeling technique for switched capacitor voltage converters
 - ▲ Voltage doubler, voltage inverter & push-pull doubler
- ▶ Average models capture circuit non-idealities (R_{ON} , R_{ESR})
- ▶ Average models may be implemented in any analog/mixed-signal HDL
- ▶ Results match circuit-level simulations faithfully
- ▶ Resulting speed gain is several thousand times

References

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- ▶ Analog Integrations Corporation. *Regulated 5V Charge Pump In SOT-23*. AIC1845.
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- ▶ Ngo, K.D.T. and Webster, R. *Steady-State Analysis and Design of a Switched-Capacitor DC-DC Converter*. Aerospace and Electronic Systems, IEEE Transactions on, Volume. 30, Issue 1, Jan. 1994.
- ▶ Sanders, S.R and Verghese, G.C. *Synthesis of Averaged Circuit Models for Switched Power Converters*, Circuits and Systems, IEEE transactions on, Volume 38, Issue 8, Aug. 1991.
- ▶ Silva-Martinez, J. *A Switched Capacitor Double Voltage Generator*.Circuits and Systems, 1994, Proceedings of the 37th Midwest Symposium on, Volume 1, Aug. 1994.
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- ▶ TianRui Ying, Wing-Hung Ki, and Mansun Chan. *Area-Efficient CMOS Charge Pumps for LCD Drivers*. Solid-State Circuits, IEEE journal of, Volume 38, Issue 10, October 2003.
- ▶ Walt Kester, Brian Erisman, Gurjit Thandi. *Section4: Switched Capacitor Voltage Converters*